

MSc on Intelligent Critical Infrastructure Systems

Machine Learning

Lecture 13: Monitoring and Control (Part II)

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funded by:

Week 1

- Introduction and Preliminaries
- Week 2
 - Linear Regression
 - Regularisation, Logistic Regression, SVMs
- Week 3
 - Neural Networks and Deep Learning
- Week 4

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- Feature Engineering and Evaluation
- Online Learning

- Week 5
 - Unsupervised Learning
- Week 6
 - Reinforcement Learning
- Week 7
 - Monitoring and Control

Application to Monitoring and Control – Outline

- Introduction to Monitoring and Control
- Mathematical Modelling of Dynamical Systems
- Adaptation and Learning in Control Systems
- System Identification (Continuous-time & Discrete-time)
- Learning Control using Neural Networks



→ CONCLUDING REMARKS FOR THE CLASS

Nonlinear Identification (Learning)

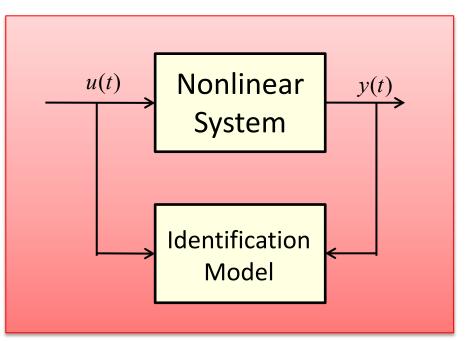
$$\dot{x}(t) = \xi(x(t), u(t), t) + f(x(t), u(t), t)$$

$$y(t) = \zeta(x(t), u(t), t) + h(x(t), u(t), t)$$

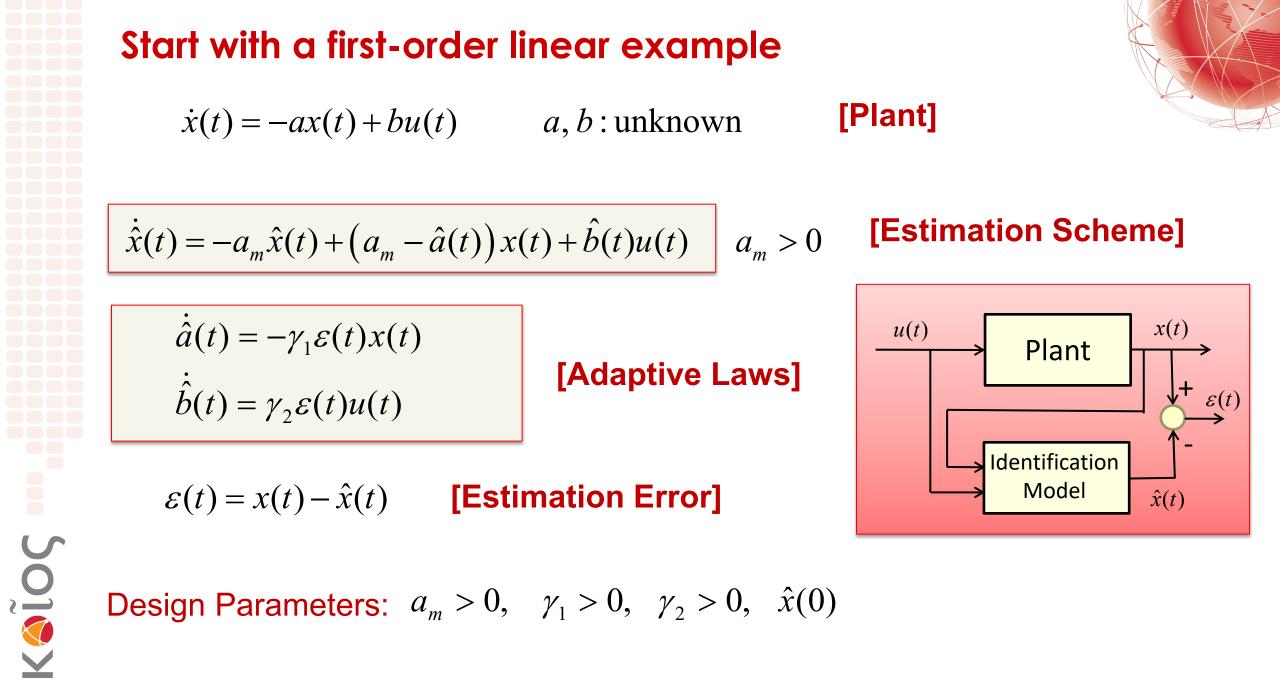
$$x(t)$$
: State variable

- u(t): Control input
- ξ, ζ : Known parts
- f, h: Unknown parts

<u>Problem:</u> design an identification model that allows estimation of the unknown f and h.







Design Parameters: $a_m > 0$, $\gamma_1 > 0$, $\gamma_2 > 0$, $\hat{x}(0)$

Nonlinear Systems with Unknown Parameters



 $\dot{x}(t) = -ax(t) + cf(x(t)) + bg(u(t))$ a, b, c : unknown $f(\bullet), g(\bullet)$: known

$$\dot{\hat{x}}(t) = -a_m \hat{x}(t) + (a_m - \hat{a}(t)) x(t) + \hat{c}(t) f(x(t)) + \hat{b}(t) g(u(t))$$
 Estimation Scheme
$$\hat{x}(t) = \frac{1}{s + a_m} \Big[(a_m - \hat{a}(t)) x(t) + \hat{c}(t) f(x(t)) + \hat{b}(t) g(u(t)) \Big]$$

Adaptive Laws

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 $\dot{\hat{a}}(t) = -\gamma_1 \varepsilon(t) x(t)$ $\dot{\hat{b}}(t) = \gamma_2 \varepsilon(t) g(u(t))$ $\dot{\hat{c}}(t) = \gamma_3 \varepsilon(t) f(x(t))$

$$\varepsilon(t) = x(t) - \hat{x}(t)$$
 [Estimation Error]

Nonlinear Systems with Unknown Functions

$$\dot{x}(t) = \underbrace{\xi(x(t), u(t))}_{\text{known}} + \underbrace{f(x(t), u(t))}_{\text{unknown}} \qquad \text{Plant}$$

$$\dot{\hat{x}}(t) = -a_m \hat{x}(t) + \xi(x(t), u(t)) + a_m x(t) + \hat{f}(x(t), u(t); \hat{\theta}(t)) \qquad \text{Estimation Scheme}$$

$$\hat{x}(t) = \frac{1}{s + a_m} \Big[\xi(x(t), u(t)) + a_m x(t) + \hat{f}(x(t), u(t); \hat{\theta}(t)) \Big]$$

Adaptive Laws

 $\dot{\hat{\theta}} = \Gamma Z^T \varepsilon$

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$$Z = \frac{\partial \hat{f}}{\partial \hat{\theta}}(x, u, \hat{\theta})$$

$$\varepsilon = x - \hat{x}$$

$$\Gamma > 0 \qquad \text{(positive definite)}$$

$$\dot{x}(t) = f(x(t)) + u(t) \qquad f : \text{unknown}$$
$$\dot{\hat{x}}(t) = -a_m \hat{x}(t) + a_m x(t) + \hat{f}(x(t); \hat{\theta}(t)) + u(t) \qquad a_m > 0$$
$$\dot{\hat{\theta}}(t) = \Gamma Z(t)^T \varepsilon(t)$$

A. Radial Basis Function (RBF) Networks

$$\hat{f}(x;\hat{\theta}_{1},\hat{\theta}_{2},\ldots\hat{\theta}_{N}) = \sum_{i=1}^{N} \hat{\theta}_{i} e^{-(x-c_{i})^{2}/\sigma^{2}}$$
$$Z_{i} = \frac{\partial \hat{f}}{\partial \hat{\theta}_{i}} = e^{-(x-c_{i})^{2}/\sigma^{2}}$$
$$\dot{\hat{\theta}}_{i} = \gamma_{i} e^{-(x-c_{i})^{2}/\sigma^{2}} (x-\hat{x})$$

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Implementation using RBF Networks

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 $x(0) = x^0$ $\hat{x}(0) = \hat{x}^0$ $\dot{x} = f(x) + u$ $\dot{\hat{x}} = -a_m \hat{x} + a_m x + \sum_{i=1}^N \left(\hat{\theta}_i e^{-(x-c_i)^2/\sigma^2}\right) + u$ $\hat{\theta}_i = \gamma_i e^{-(x-c_i)^2/\sigma^2} \left(x - \hat{x}\right)$ $\hat{\theta}_i(0) = \hat{\theta}_i^0$ Design parameters to choose:

$$, \sigma, \hat{x}^0, \gamma_i, c_i, \theta_i^0 \qquad \qquad i = 1, 2, \dots N$$

B. Sigmoidal Neural Networks (SNN)

$$\hat{f}(x;\hat{\theta}_{1},\hat{\theta}_{2},\ldots\hat{\theta}_{3M}) = \sum_{i=1}^{M} \hat{\theta}_{i} \sigma\left(\hat{\theta}_{i+M}x + \hat{\theta}_{i+2M}\right)$$
$$\sigma\left(p\right) = \frac{1}{1+e^{-p}} \implies \sigma\left(\alpha x + \beta\right) = \frac{1}{1+e^{-(\alpha x + \beta)}}$$

$$Z_{i} = \frac{\partial \hat{f}}{\partial \hat{\theta}_{i}}$$
$$\dot{\hat{\theta}}_{i} = \gamma_{i} Z_{i} \left(x - \hat{x} \right)$$

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Implementation of SNN

$$\dot{x} = f(x) + u \qquad x(0) = x^{0}$$

$$\dot{\hat{x}} = -a_{m}\hat{x} + a_{m}x + \sum_{i=1}^{M}\hat{\theta}_{i}\sigma\left(\hat{\theta}_{i+M}x + \hat{\theta}_{i+2M}\right) + u \qquad \hat{x}(0) = \hat{x}^{0}$$

$$\dot{\hat{\theta}}_{i} = \gamma_{i}\frac{\partial\hat{f}}{\partial\hat{\theta}_{i}}(x - \hat{x}) \qquad \hat{\theta}_{i}(0) = \hat{\theta}_{i}^{0}$$

Design parameters to choose:

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 $a_m, \hat{x}^0, \gamma_i, \hat{\theta}_i^0 \qquad \qquad i = 1, 2, \dots 3M$

Discrete-time Learning in dynamical systems

$$y(k+1) = f(y(k), y(k-1), \dots, y(k-n_y, u(k), \dots, u(k-n_u)))$$

$$k \in \mathbb{Z}^{+} \qquad k = 0, 1, 2, \dots$$

$$f : \mathbb{R}^{n_{y}+1} \times \mathbb{R}^{n_{u}+1} \to \mathbb{R}$$

$$f \quad \text{is unknown (or partially unknown)}$$
Let
$$z(k) = \left[y(k), y(k-1), \dots y(k-n_{y}, u(k), \dots u(k-n_{u}) \right]$$

$$\Rightarrow y(k+1) = f(z(k)) \qquad (z(k) \text{ is measurable})$$

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Discrete-Time Learning

$$y(k+1) = f(z(k))$$

$$\hat{y}(k+1) = \hat{f}(z(k); \hat{\theta}(k))$$
Prediction/Identification Model
Adaptive Laws:

$$\hat{\theta}(k+1) = \hat{\theta}(k) + \alpha_0 e(k+1)\xi(k)$$

$$\alpha_0 > 0$$

$$\alpha_0 > 0$$

$$\beta(k+1) = \hat{\theta}(k) + \alpha_0 e(k+1)\xi(k)$$

$$\alpha_0 > 0$$

$$\beta(k+1) = \hat{\theta}(k) + \frac{\gamma_0 e(k+1)}{\beta_0 + |\xi(k)|^2}\xi(k)$$

$$\hat{\theta}(k+1) = \hat{\theta}(k) + \frac{\gamma_0 e(k+1)}{\beta_0 + |\xi(k)|^2}\xi(k)$$

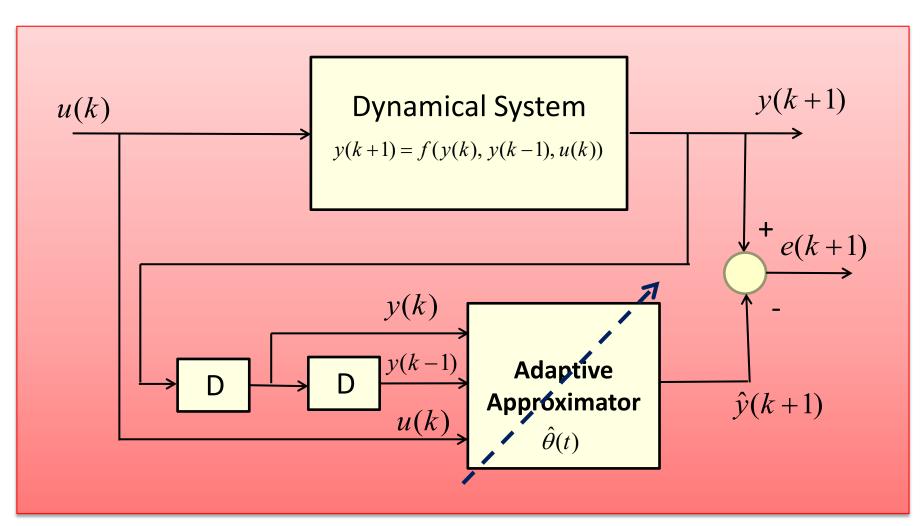
$$\hat{\xi}(k) = \frac{\partial \hat{f}}{\partial \hat{\theta}}(z(k), \hat{\theta}(k))$$
Normalized Gradient Descent
$$e(k) = y(k) - \hat{y}(k)$$
Estimation error

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Discrete-Time Learning

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Example: y(k+1) = f(y(k), y(k-1), u(k))





Discrete-Time Learning

Example: y(k+1) = f(y(k), y(k-1), u(k))Given $y(0), y(-1), \hat{\theta}(0), u(0)$ \begin {for} $k = 0, 1, 2, \dots N$ 1. z(k) = [y(k), y(k-1), u(k)]2. $\xi(k) = \frac{\partial \hat{f}}{\partial \hat{\theta}} (z(k), \hat{\theta}(k))$ 3. y(k+1) = f(z(k))4. $\hat{v}(k+1) = \hat{f}(z(k), \hat{\theta}(k))$ 5. $\hat{\theta}(k+1) = \hat{\theta}(k) + \frac{\gamma_0 \left(y(k+1) - \hat{y}(k+1) \right)}{\beta_0 + \left| \xi(k) \right|^2} \xi(k)$

 $end {for}$

From Learning to Fault Diagnosis

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- A similar estimation scheme can also be used for fault diagnosis
- Additionally, in fault diagnosis we require upper bounds (or statistical information) on the uncertainty so that we can distinguish between faults and modelling uncertainties
- Based on the modelling uncertainty bounds, we can compute adaptive threshold signals
- Finally, we also need to design a decision logic algorithm for fault detection and isolation

Learning Control Using Neural Networks

$$\dot{x}(t) = \xi(x(t), u(t), t) + f(x(t), u(t), t)$$

$$y(t) = \zeta(x(t), u(t), t) + h(x(t), u(t), t)$$

x(t): State variable

- u(t): Control input
- ξ, ζ : Known parts
- f, h: Unknown parts

Problem: design a controller:

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 $u = \mu(y, \theta, t)$ $\dot{\theta} = \lambda(u, y, \theta, t)$

Such that y(t) follows a desired trajectory $y_r(t)$ as closely as possible.

Start with a simple case: linear adaptive control



 $\dot{y}(t) = ay(t) + bu(t)$ a, b: unknown; sign(b) is known

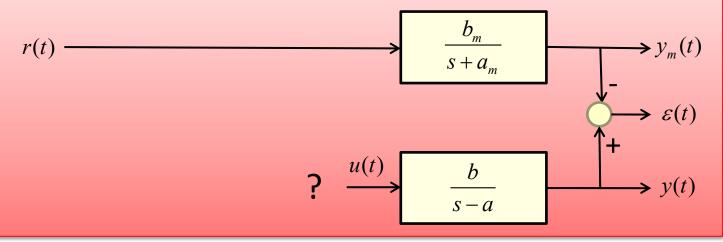
Model Reference Adaptive Control (MRAC) Problem:

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Choose an appropriate control law u(t) such that all signals in the closed loop plant are bounded and the plant output y(t) tracks the output $y_m(t)$ of the reference model:

$$\dot{y}_m(t) = -a_m y_m(t) + b_m r(t)$$

where r(t) is a bounded, piecewise continuous signal, referred to as the reference or command input.



Start with a simple case: linear adaptive control

How would we solve the MRAC problem if we knew a, b?

 $u = -k^* y(t) + l^* r(t)$

$$\Rightarrow k^* = \frac{a_m + a}{b} \qquad l^* = \frac{b_m}{b}$$

Assuming that b is not equal to 0 (plant is controllable)

$$\frac{y(s)}{r(s)} = \frac{y_m(s)}{r(s)} \implies |y(t) - y_m(t)| \to 0 \text{ (exponentially fast for all r(t))}$$



Direct Adaptive Control (Direct MRAC)

$$\dot{y}(t) = ay(t) + bu(t)$$

$$\dot{y}_m(t) = -a_m y_m(t) + b_m r(t)$$

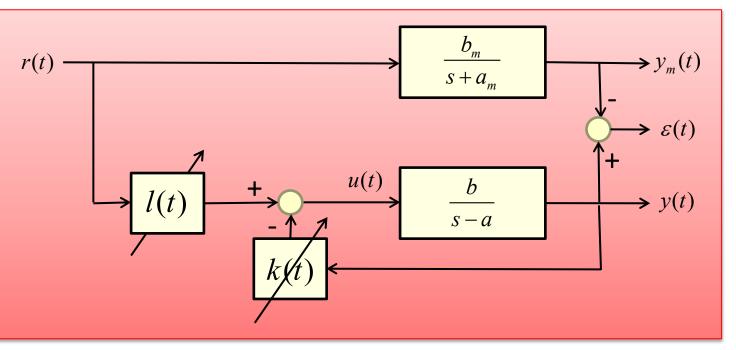
$$u(t) = -k(t)y(t) + l(t)r(t)$$

$$\varepsilon(t) = y(t) - y_m(t)$$

 $\dot{y}(t) = (a - bk(t))y(t) + bl(t)r(t)$ $\dot{y}_m(t) = -a_m y_m(t) + b_m r(t)$

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$$\dot{k}(t) = \gamma_1 \varepsilon(t) y(t) \operatorname{sgn}(b)$$
$$\dot{l}(t) = -\gamma_2 \varepsilon(t) r(t) \operatorname{sgn}(b)$$



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Indirect Adaptive Control (Indirect MRAC)

If a, b were known, then we would use:

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$$u = -k^* y(t) + l^* r(t)$$

$$\begin{cases} k^* = \frac{a_m + a}{b} \\ l^* = \frac{b_m}{b} \end{cases}$$

Replace the unknown a, b by their estimates $\hat{a}(t), \hat{b}(t)$ where they are generated by identification techniques:

$$u(t) = -\frac{a_m + \hat{a}(t)}{\hat{b}(t)} y(t) + \frac{b_m}{\hat{b}(t)} r(t)$$
(IAC)
$$u(t) = -k(t)y(t) + l(t)r(t)$$
(DAC)

Need to make sure that: $\hat{b}(t) \neq 0$ (otherwise $u(t) \rightarrow \infty$)

Indirect Adaptive Control (Indirect MRAC)

Identification Model (Estimation Scheme):

 $\dot{\hat{y}}(t) = -a_m \hat{y}(t) + (a_m + \hat{a}(t))y(t) + \hat{b}(t)u(t)$

Identification Error:

$$\mathcal{E}_i(t) = y(t) - \hat{y}(t)$$

Adaptive Laws:

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$$\dot{\hat{a}}(t) = \gamma_1 \varepsilon_i(t) y(t)$$
$$\dot{\hat{b}}(t) = \gamma_2 \varepsilon_i(t) u(t)$$



Indirect Adaptive Control (Indirect MRAC)

How do we guarantee that: $\hat{b}(t) \neq 0$

Instead of:
$$\hat{b}(t) = \gamma_2 \varepsilon_i(t) u(t)$$

Use:

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$$\dot{\hat{b}}(t) = \begin{cases} \gamma_2 \varepsilon_i(t) u(t) & \text{if } \hat{b} > \overline{b} \\ \gamma_2 \varepsilon_i(t) u(t) & \text{if } \hat{b} = \overline{b} \text{ and } \gamma_2 \varepsilon_i u > 0 \\ 0 & \text{if } \hat{b} = \overline{b} \text{ and } \gamma_2 \varepsilon_i u \le 0 \end{cases}$$

This will guarantee that
$$\hat{b}(t) \ge \overline{b}$$
 and also $u \in L_{\infty}$

Need to assume that we know the sign of b and also that we know a lower bound on b.

Nonlinear Systems (with unknown functions)

$$\dot{y} = f(y) + u$$

Objective: Choose an appropriate control law u(t) such that all the signals in the closed-loop plant are bounded and the plant output y(t) tracks the output $y_m(t)$ of the reference model

$$\dot{y}_m = -a_m y_m + b_m r$$

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for all r(t) that are bounded and piecewise continuous.

Nonlinear Systems (with unknown functions)

If we knew f(y) then what would we do?

$$u = -f(y) - a_m y + b_m r$$

$$\dot{y} = f(y) - f(y) - a_m y + b_m r$$

$$\dot{y} = -a_m y + b_m r$$

$$\dot{y}_m = -a_m y_m + b_m r$$

$$\lim_{t \to \infty} \left\| y(t) - y_m(t) \right\| = 0$$

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Exponentially fast

Learning Control with Neural Networks (indirect approach)

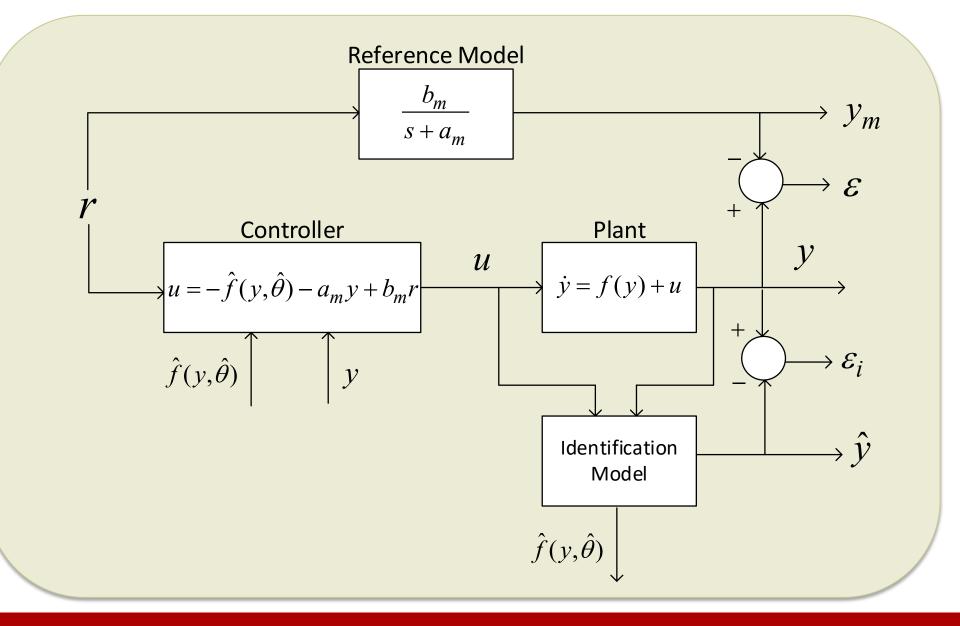
$$u = -\hat{f}(y,\hat{\theta}) - a_m y + b_m r$$

Identification Model (Estimation Scheme):

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$$\begin{split} \dot{\hat{y}} &= -a_m \hat{y} + a_m y + \hat{f}(y, \hat{\theta}) + u \\ \dot{\hat{\theta}} &= \Gamma Z^\top e_i \\ & Z = \frac{\partial \hat{f}(\hat{y}, \hat{\theta})}{\partial \hat{\theta}} \end{split}$$

Learning Control with Neural Networks (indirect approach)



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Learning Control with Neural Networks (indirect approach)

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$$\begin{split} \dot{y} &= f(y) \underbrace{-\hat{f}(y,\hat{\theta}) - a_m y + b_m r}_{u}, \quad y(0) = y_o \\ \dot{y} &= -a_m \hat{y} + a_m y + \hat{f}(y,\hat{\theta}) + u, \quad \hat{y}(0) = \hat{y}_o \\ \dot{y}_m &= -a_m y + b_m r, \quad y_m(0) = y_m^o \\ \dot{\theta} &= \Gamma \frac{\partial \hat{f}}{\partial \hat{\theta}} \underbrace{(y - \hat{y})}_{e_i}, \quad \hat{\theta}(0) = \hat{\theta}^o \end{split}$$

Discrete-time Learning Control

$$y(k+1) = f(y(k), y(k-1), \dots y(k-n_y) + u(k)$$

Objective: Choose an appropriate control law u(k) such that all the signals in the closed-loop plant are bounded and the plant output y(k) tracks the output $y_m(k)$ of the reference model $y_m(k+1) = -a_m y_m(k) + b_m r(k)$ for all r(k) that are bounded.

f is unknown (or partially unknown)

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Let
$$z(k) = [y(k), y(k-1), \dots y(k-n_y)]$$

 $\Rightarrow y(k+1) = f(z(k)) + u(k)$ (z(k) is measurable)

Discrete-Time Learning Control

$$y(k+1) = f(z(k)) + u(k)$$

$$\hat{y}(k+1) = \hat{f}(z(k); \hat{\theta}(k)) + u(k) \quad \longleftarrow \quad \text{Prediction/Identification Model}$$

$$u(k) = -\hat{f}(z(k); \hat{\theta}(k)) - a_m y_m(k) + b_m r(k) \quad \longleftarrow \quad \text{Learning Control Law}$$

Adaptive Laws:

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$$\hat{\theta}(k+1) = \hat{\theta}(k) + \alpha_0 e(k+1)\xi(k)$$

$$\hat{\theta}(k+1) = \hat{\theta}(k) + \frac{\gamma_0 e(k+1)}{\beta_0 + \left|\xi(k)\right|^2} \xi(k)$$

Normalized Gradient Descent

 $\begin{aligned} \alpha_0 &> 0 & \text{Step size} \\ 0 &< \gamma_0 &< 2 & \text{Learning rate} \\ \beta_0 &> 0 & \text{Small design constant} \\ \xi(k) &= \frac{\partial \hat{f}}{\partial \hat{\theta}} \Big(z(k), \hat{\theta}(k) \Big) \frac{\text{Network sensitivity}}{\text{function}} \\ e(k) &= y(k) - \hat{y}(k) & \text{Estimation error} \end{aligned}$



Discrete-Time Learning Control

Example: y(k+1) = f(y(k), y(k-1)) + u(k)

Given $y(0), y(-1), \hat{\theta}(0), y_m(0)$ \begin {for} $k = 0, 1, 2, \dots, N$ 1. z(k) = [y(k), y(k-1)]2. $u(k) = -\hat{f}(z(k), \hat{\theta}(k)) - a_m y_m(k) + b_m r(k)$ 3. $\xi(k) = \frac{\partial \hat{f}}{\partial \hat{\theta}} (z(k), \hat{\theta}(k))$ 4. y(k+1) = f(z(k)) + u(k)5. $y_m(k+1) = -a_m y_m(k) + b_m r(k)$ 6. $\hat{v}(k+1) = \hat{f}(z(k), \hat{\theta}(k))$ 7. $\hat{\theta}(k+1) = \hat{\theta}(k) + \frac{\gamma_0 \left(y(k+1) - \hat{y}(k+1)\right)}{\beta_0 + \left|\xi(k)\right|^2} \xi(k)$

 $\end {for}$

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Some Bibliography on Monitoring and Control Using Learning Methods

- J.A. Farrell and M.M. Polycarpou, "Adaptive Approximation Based Control", J. Wiley, 2006.
- M. Blanke, M. Kinnaert, J. Lunze, and M. Staroswiecki.
 Diagnosis and Fault-Tolerant Control. Springer-Verlag, 2016.

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Solution]



Some Final Conclusions

- We have learned a lot
- > A lot that we have not learned!
- > The importance of gaining intuition
- Never lose the big picture and the applications